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## Design Note – DN05037/D

# **Extremely Low Power Off-line Supply**

Device	Application	Input Voltage	<b>Output Power</b>	Topology	I/O Isolation
NCP1050	Sensors	90 - 300 Vac	15 mW	Flyback	None

## **Other Specifications**

Output Voltage	5 Vdc	
Nominal Current	3 mA	
Input Consumption	<0.5 mArms	

### Introduction

There are many power solutions that comply with the Energy Star directives for stand-by / no load consumption. Most of them rely on converters that switch in bursts or get into skipping mode.

The NCP1050 is a good candidate for such applications and this design note describes a simple off-line power supply that operates at a very low power consumption. It is used to power up a motion / infrared sensor and the logic circuit associated with it. This supply provides 15 mW output power while consuming only 0.5 mA from the AC line in order not to trip the GFCI protection.

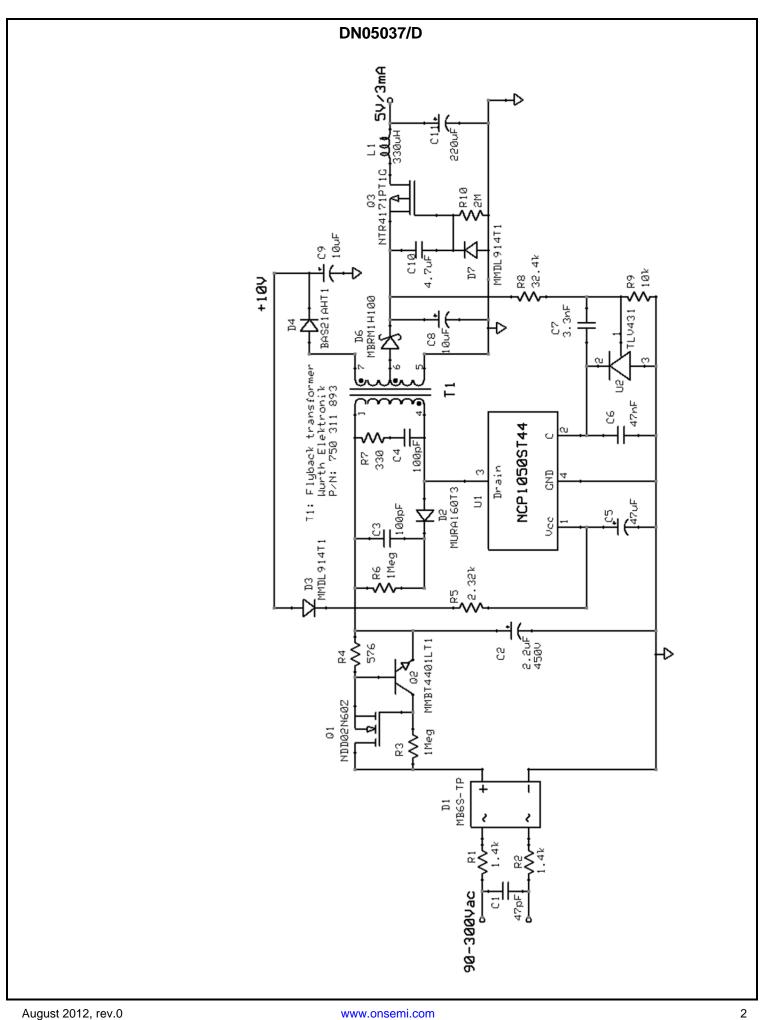
## **Circuit Description**

The NCP105X family of gated oscillator type converters is very suitable for this application. In order to minimize losses the NCP1050 is chosen because it has the lowest peak current level in the family and provides 44 kHz operation.

The circuit is a typical flyback and since no isolation is required it is pretty simple. The secondary has basically two windings: one provides the main output (5 V) and the other one provides the bias (10 V) for the NCP1050. Although the NCP1050 can be self biased, using an extra winding for bias results in even lower power consumption.

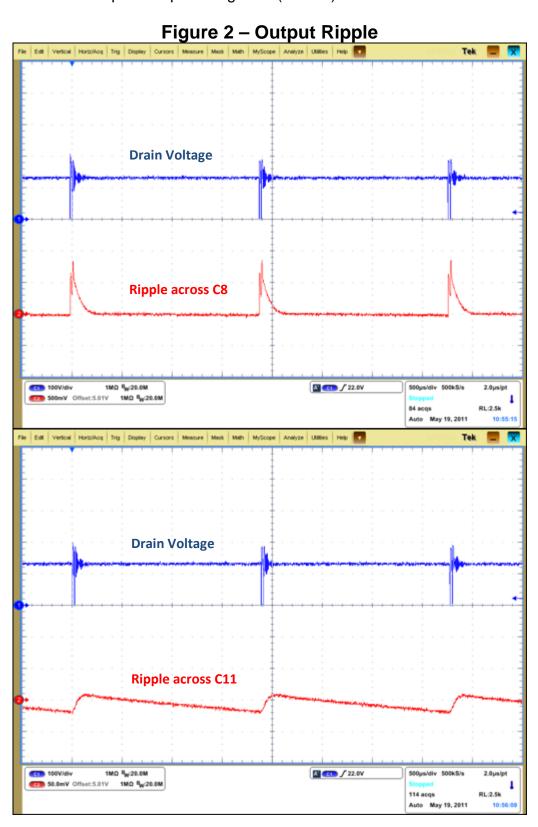
The input filtering capacitor (C2) needs to have a very low value in order to minimize the rms input current. A 2.2 uF value works very well. To reduce even more the current spikes due to the input capacitor being charged each cycle, a current limiter (Q1 - Q2) is also added. With all these precautions the converter produces 5 V / 3 mA output while consuming 0.49 mA at 110 Vac while maintaining the inrush current below 2 mA. Input power at 110 Vac input was 51 mW (measured with Yokogawa WT210 power meter and averaged over a 2 minute period).

# Figure 1 - Schematic



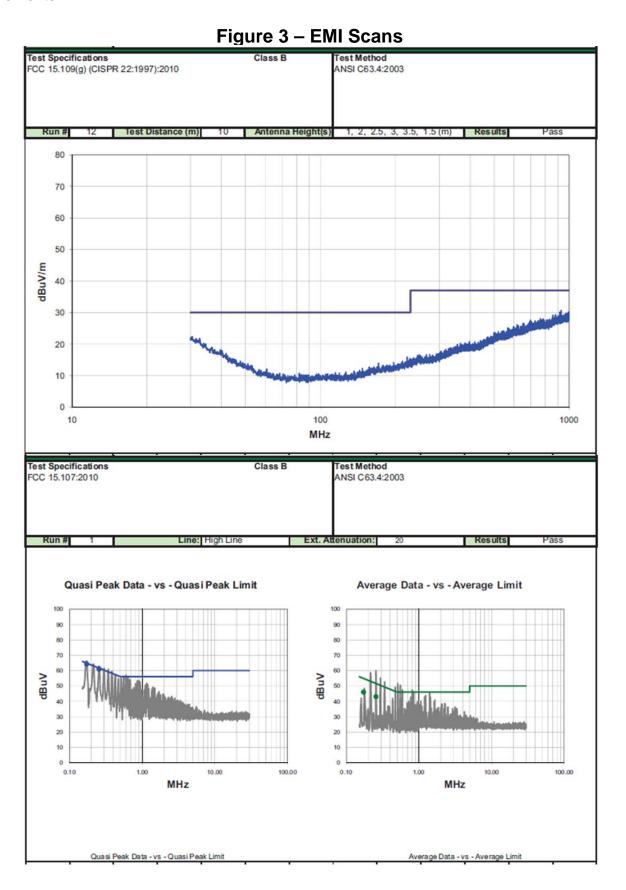
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Due to the fact that power is delivered in bursts, the output ripple is significant. Figure 2 (top) shows almost 0.8 V peak voltage ripple across capacitor C8. Beefing up C8 is not a viable option because, in this case, the input current limiter (Q1 - Q2) would prevent it from being charged and the converter will enter a hiccup mode, and will never be able to fully start. Therefore a series transistor (Q3) is added after C8 to act like a soft start and slowly charge a larger capacitor C11. This way, voltage ripple across C11 is reduced bellow S0 mV peak-to-peak - figure 2 (bottom)



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Resistors R1 – R2, in front of the bridge, also help reduce current spikes and, in conjunction with capacitor C1, act like an input EMI filter. The EMI scans in figure 3 show the circuit passes Class B requirements.



## DN05037/D Figure 4 – Input current at 110 Vac input

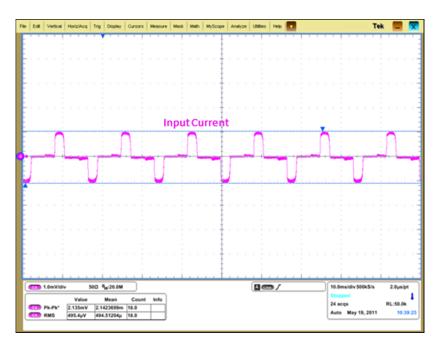
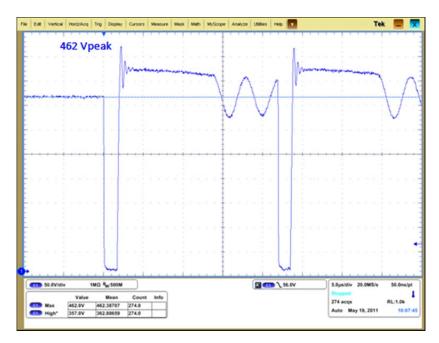
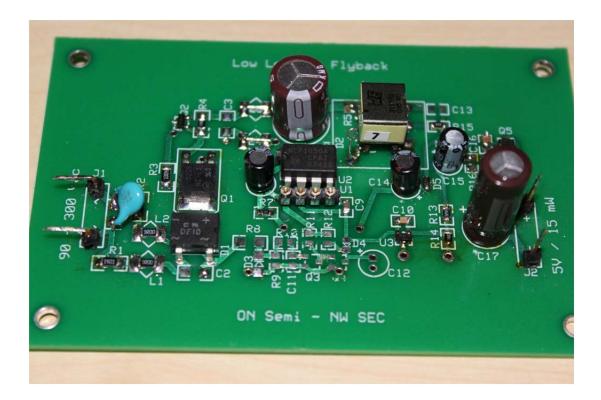


Figure 5 - Maximum drain voltage at 300 Vac input



### **Board**



## **Key Features**

- Simple topology
- Low part count
- Extremely low consumption
- Wide input range

### References

Data sheet NCP1050: Monolithic High Voltage Gated Oscillator Power Switching Regulator

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