CCM Buck Controller for Precise Current Regulation and Wide Analog Dimming

NCL35076

The NCL35076 is a DC–DC buck controller for wide dimming range down to 1% by analog dimming control to relieve audible noise and flicker in PWM dimming. ON Semiconductor's proprietary LED current calculation technique driven by zero input offset amplifiers performs precise constant current in the whole analog dimming range. Multi–mode operation provides low LED current ripple with small output capacitor by CCM at heavy load and deep analog dimming by DCM at light load.

PWM dimming is also provided in case that constant LED color temperature is required. NCL35076 ensures high system reliability with LED short protection, over current protection and thermal shutdown.

Features

- Wide Analog Dimming Range: 1~100%
- Low CC Tolerance: ±2% at 100% Load & ±20% at 1% Load
- Low System BOM
- LED Off Mode at Standby
- Low Standby Current
- PWM Dimming Available
- Gate Sourcing and Sinking Current of 0.5 A/0.8 A
- Robust Protection Features
 - LED Short Protection
 - Over Current Protection
 - Thermal Shutdown
 - ♦ V_{DD} Over Voltage Protection

Typical Applications

• LED Lighting System



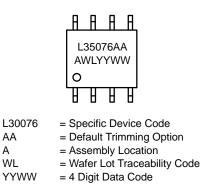
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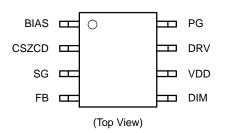


SOIC-8 NB CASE 751

MARKING DIAGRAM



PIN ASSIGNMENT



ORDERING INFORMATION

| Device | Package | Shipping |
|----------------|-----------|-----------------------|
| NCL35076AADR2G | SOIC-8 NB | 3000 / Tape & Reel |

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

APPLICATION SCHEMATIC

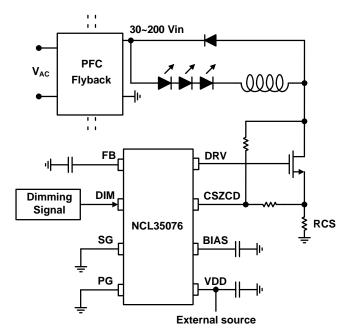


Figure 1. Application Schematic

BLOCK DIAGRAM

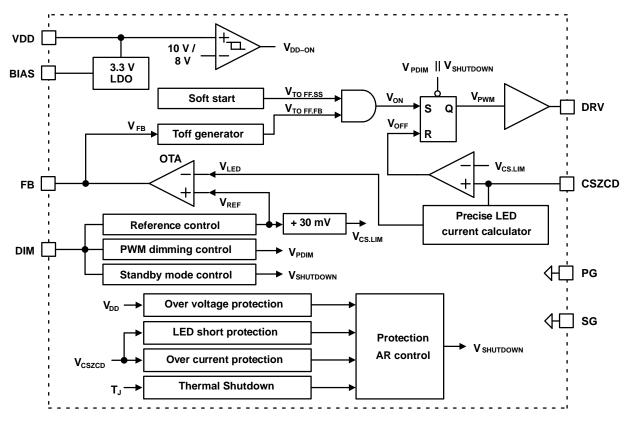


Figure 2. Simplified Block Diagram

PIN CONFIGURATION

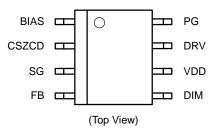


Figure 3. Pin Configuration

PIN FUNCTION DESCRIPTION

| Pin No. | Pin Name | Function | Description |
|---------|----------|--------------------|--|
| 1 | BIAS | 3.3 V BIAS | This pin is 3.3 V LDO output to bias the internal digital circuit |
| 2 | CSZCD | CS and ZCD Sensing | This pin detects the switch current and the inductor current zero cross time |
| 3 | SG | Signal Ground | Signal Ground is close to control pin circuit such as CSZCD, DIM and FB |
| 4 | FB | Feedback | Output of feedback OTA |
| 5 | DIM | Dimming Input | Dimming signal is provided to this pin |
| 6 | VDD | Power Supply | IC operating current is supplied to this pin |
| 7 | DRV | Output Drive | This pin is connected to drive external switch |
| 8 | PG | Power Ground | Power Ground is close to the capacitors at BIAS and VDD pin |

SPECIFICATIONS

MAXIMUM RATINGS

| Parameter | Symbol | Value | Unit |
|---|-----------------------|-------------|------|
| VDD, DRV Pin Voltage Range | V _{MV(MAX)} | -0.3 to 30 | V |
| DIM, FB, CSZCD, BIAS Pin Voltage Range | V _{LV(MAX)} | -0.3 to 5.5 | V |
| Maximum Power Dissipation ($T_A < 50^{\circ}C$) | P _{D(MAX)} | 550 | mW |
| Maximum Junction Temperature | T _{J(max)} | 150 | °C |
| Storage Temperature Range | T _{STG} | -55 to 150 | °C |
| Junction-to-Ambient Thermal Impedance | $R_{	extsf{	heta}JA}$ | 145 | °C/W |
| ESD Capability, Human Body Model (Note 2) | ESD _{HBM} | 2 | kV |
| ESD Capability, Charged Device Model (Note 2) | ESD _{CDM} | 1 | kV |

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTICS, RECOMMENDED OPERATING RANGES and/or APPLICATION INFORMATION for Safe Operating parameters.

2. This device series incorporates ESD protection and is tested by the following methods:

– ESD Human Body Model per JEDEC Standard JESD22–A114
 – ESD Charged Device Model per JEDEC Standard JESD22–C101

- Latch-up Current Maximum Rating ±100 mA per JEDEC Standard JESD78

RECOMMENDED OPERATING RANGES

| Parameter | Symbol | Min | Max | Unit |
|----------------------|--------|-----|-----|------|
| Junction Temperature | TJ | -40 | 125 | °C |

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

ELECTRICAL CHARACTERISTICS (V_{DD} = 15 V and T_J = -40~125°C unless otherwise specified)

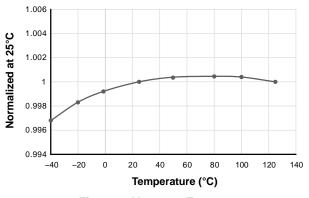
| Parameter | Test Conditions | Symbol | Min | Тур | Max | Unit |
|---------------------------------------|---|---------------------------|-------|-------|-------|------|
| VDD SECTION | - | | | | | |
| IC Turn–On Threshold Voltage | | V _{DD(ON)} | 9.3 | 10.0 | 10.7 | V |
| IC Turn–Off Threshold Voltage | | V _{DD(OFF)} | 7.4 | 8.0 | 8.6 | V |
| Startup Current | $V_{DD} = V_{DD(ON)} - 1.6 \text{ V}$ | I _{DD(ST)} | - | 250 | 400 | μΑ |
| Operating Current | | I _{DD(OP)} | - | 6.5 | 8.0 | mA |
| Standby Current | | I _{DD(SB)} | - | 200 | 300 | μΑ |
| BIAS SECTION | | | | | | |
| BIAS Voltage | | V _{BIAS} | 3.23 | 3.30 | 3.37 | V |
| | T _J = 25~100°C (Note 4) | | 3.25 | 3.30 | 3.35 | V |
| DIM SECTION | | | | | | |
| DIM Voltage for 100% V _{REF} | V _{DIM} = 2.6 V | V _{DIM(REF-MAX)} | 2.44 | 2.50 | 2.56 | V |
| DIM Voltage for 99% V _{REF} | | V _{DIM(MAX-EFF)} | 2.400 | 2.475 | 2.528 | V |
| Standby Enabling DIM Voltage | | V _{DIM(SB-ENA)} | 50 | 75 | 100 | mV |
| Standby Disabling DIM Voltage | | V _{DIM(SB-DIS)} | 60 | 100 | 140 | mV |
| Standby Delay Time | | t _{SB(DELAY)} | 9 | 10 | 11 | ms |
| FB SECTION | | | | | | |
| FB OTA Source Current | $\label{eq:IFB} \begin{array}{l} IFB = (V_{LED} - V_{REF}) \ x \ g_{M(FB)} \ x \ 12.5 \\ V_{REF} = 120 \ mV, \ V_{LED} = 80 \ mV \end{array}$ | I _{FB(SOURCE)} | -14.0 | -11.5 | -9.0 | μΑ |
| FB OTA Sink Current | $\label{eq:IFB} \begin{array}{l} IFB = (V_{LED} - V_{REF}) \ x \ g_{M(FB)} \ x \ 12.5 \\ V_{REF} = 40 \ mV, \ V_{LED} = 80 \ mV \end{array}$ | I _{FB(SINK)} | 9.0 | 11.5 | 14.0 | μΑ |

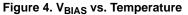
| Parameter | Test Conditions | Symbol | Min | Тур | Max | Unit |
|--|--|---------------------------|------|------|------|------|
| FB SECTION | | • | | | | |
| FB OTA Transconductance | $g_{M(FB)} = I_{FB} / \{(V_{REF} - V_{LED}) \times 12.5\}$ | 9 _{М(FB)} | 18 | 23 | 28 | μmho |
| FB OTA High Voltage | V_{REF} = 120 mV, V_{LED} = 80 mV | V _{FB(HIGH)} | 4.7 | - | - | V |
| FB Minimum Clamping Voltage | $V_{REF} = 0 \text{ mV}, V_{LED} = 80 \text{ mV}$ | V _{FB(CLP)} | 0.4 | 0.5 | 0.6 | V |
| CS SECTION | - | | | | | |
| CS Regulation | | V _{CS(REG-MAX)} | 175 | 180 | 185 | mV |
| CS Current Ripple Voltage | | V _{CS(RIPPLE)} | 25 | 30 | 35 | mV |
| CS Current Limit Minimum | | V _{CS(LIM-MIN)} | 72 | 80 | 88 | mV |
| DUTY SECTION | | | | | | - |
| Leading Edge Blanking Time at Turn–on | | t _{LEB(TON)} | 360 | 400 | 440 | ns |
| Maximum Ton Time | | t _{ON(MAX)} | 45 | 50 | 55 | μs |
| Minimum Toff Time | V _{FB} = 3.8 V | t _{OFF(MIN)} | 400 | 850 | 1000 | ns |
| Maximum Toff Time | V _{FB} = 0.5 V | t _{OFF(MAX)} | 1.17 | 1.30 | 1.43 | ms |
| Maximum FB Voltage for Min. Toff | | V _{FB(MAX-TOFF)} | 3.30 | 3.43 | 3.55 | V |
| Minimum FB Voltage for Max. Toff | | V _{FB(MIN-TOFF)} | 0.9 | 1.1 | 1.3 | V |
| DRV SECTION | | | | | | |
| DRV Low Voltage | | V _{DRV(LOW)} | _ | - | 0.2 | V |
| DRV High Voltage | V _{DD} = 15 V | V _{DRV(HIGH)} | 11 | 12 | 13 | V |
| DRV Rising Time | C _{DRV} = 3.3 nF | t _{DRV(R)} | 60 | 100 | 145 | ns |
| DRV Falling Time | C _{DRV} = 3.3 nF | t _{DRV(F)} | 25 | 55 | 105 | ns |
| AUTO RESTART SECTION | | | | | | |
| Auto Restart Time at Protection | | t _{AR(PROT)} | 0.9 | 1.0 | 1.1 | S |
| VDD OVER VOLTAGE PROTECTIO | N SECTION | | | | | |
| VDD Over Voltage Threshold Voltage | 9 | V _{DD(OVP)} | 22 | 23 | 24 | V |
| SHORT LED PROTECTION SECTIO | DN | | | | | |
| SLP Monitoring Triggering Delay Time | | t _{SLP(MON-DEL)} | 18 | 20 | 22 | ms |
| SLP Monitoring Disable Time at Startup | | t _{SLP(MON-DIS)} | 10.8 | 12.0 | 13.2 | ms |
| OVER CURRENT PROTECTION SE | CTION | | | | | |
| CS Over Current Protection Threshold | | V _{CS(OCP)} | 0.4 | 0.5 | 0.6 | V |
| THERMAL SHUTDOWN SECTION | | | | | | |
| Thermal Shut Down Temperature (Note 3) | | T _{SD} | 130 | 150 | 170 | °C |
| Thermal Shut Down Hysteresis (Note 3) | | T _{SD(HYS)} | 25 | 30 | 35 | °C |

(Note 3)
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.
3. Guaranteed by design.
4. Guaranteed by characterization.

TYPICAL CHARACTERISTICS

(These characteristic graphs are normalized at T_{A} = 25°C)





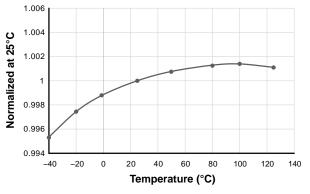


Figure 5. V_{DIM(MAX)} vs. Temperature

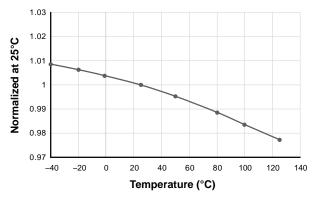


Figure 6. $g_{M(FB)}$ vs. Temperature

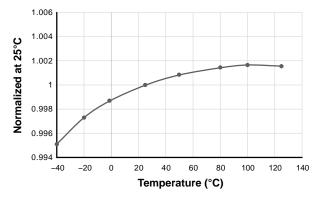
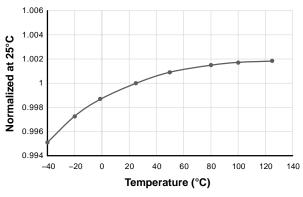


Figure 7. V_{CS(REG-MAX)} vs. Temperature





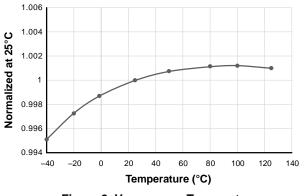


Figure 9. V_{DD(OVP)} vs. Temperature

APPLICATION INFORMATION

General

NCL35076 provides wide analog dimming down to 1% with high CC accuracy. According to buck inductor, input voltage and output voltage, deep dimming down to 0.1~0.2% load can be achieved. Thanks to the ON semiconductor's proprietary LED current calculation technique, NCL35076 is able to measure the current of LED load connected at input voltage node without the upper limit of the input voltage with high system reliability. LED current is sensed and regulated by internal zero input offset amplifiers so that NCL35076 performs precise CC regulation in the whole analog dimming range. Therefore, CC tolerance is tightly controlled in $\pm 2\%$ at 100% load and $\pm 20\%$ at 1% load.

Wide Analog Dimming

Wide analog dimming range is obtained by transitioning multi-mode between CCM and DCM according to the dimming condition. At full load condition, CCM with $\pm 17\%$ inductor current ripple minimizes the conduction loss with high efficiency and DCM is entered at light load condition to perform analog deep dimming. Dimming curve linearity is obtained by a digital compensator in the entire dimming range.

PWM Dimming

Analog dimming has benefits for less audible noise and flicker compared to PWM dimming. However, there is a need of PWM dimming method to keep the constant LED color temperature in specific applications. NCL35076 supports PWM dimming by simply inputting PWM dimming signal to DIM pin.

Precise CC Regulation

CC regulation is very important especially in programmable LED driver because the driver should keep precise CC control under the system variation of LED load, inductor, temperature, etc. Since NCL35076 applies zero input offset amplifiers at LED current calculator block and OTA, CC tolerance is less than $\pm 2\%$ at 100% load and $\pm 20\%$ at 1% load in the system variation.

Soft start

Without soft start in the closed loop CC control, the LED current overshoot is easily occurred at startup so that the overshoot can affect a lifetime of LEDs and incur an undesirable flash. NCL35076 provides soft start technique to prevent the LED current overshoot by T_{OFF} time control.

Standby Mode

When V_{DIM} is lower than a standby threshold voltage for 10 ms, standby mode is triggered with LED turn–off and IC current consumption is minimized.

Auto Restart (AR) at Protection

Once protection is triggered, IC operation stops for 1 second and begins soft start operation after the auto restart time delay.

VDD Over Voltage Protection (OVP)

When VDD is higher than 23 V, over voltage protection is triggered.

Short LED Protection (SLP)

When LED is short circuited, the buck stage operates in CCM with maximum turn-off time. By detecting this condition, short LED protection is triggered.

Over Current Protection (OCP)

When CSZCD voltage exceeds the over current threshold voltage, switching is immediately shut down after leading edge blanking time in the short circuit condition of the inductor, the freewheeling diode or the LED load.

Thermal Shot Down (TSD)

When IC junction temperature is higher than 150°C, TSD is triggered and released when the temperature is lower than 120°C.

BASIC OPERATION

NCL35076 is the current mode buck controller in which DRV is off when V_{CSZCD} reaches to $V_{CS.LIM}$ (= V_{REF} + 30 mV) and DRV is on by T_{OFF} generator controlled by V_{FB} . V_{LED} is calculated based on V_{CSZCD} in precise LED current calculator block composed of zero input offset amplifiers and V_{REF} is controlled by DIM signal. In reference control block, V_{REF} is obtained by below equation.

$$V_{\text{REF}}[V] = \frac{V_{\text{DIM}} - 0.25 V}{12.5}$$
 (eq. 1)

 V_{LED} is compared with V_{REF} at OTA to generate V_{FB} which controls T_{OFF} time in T_{OFF} generator. T_{OFF} is inversely proportional to V_{FB} . Therefore, T_{OFF} is shorter as V_{FB} increases. T_{OFF} is set by below equation.

$$T_{OFF} [\mu s] = \frac{1.8}{V_{FB} - 1.1} + 0.1$$
 (eq. 2)

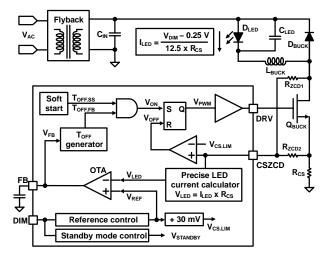


Figure 10. NCL35076 Block Diagram

Wide Analog Dimming

NCL35076 operates in CCM at heavy load and in DCM at light load for a wide analog dimming. Figure 11 shows how NCL35076 operates with V_{DIM} .

- <u>A:</u> $V_{CS,LIM}$ follows V_{REF} + 30 mV which is ±17% inductor current ripple at 2.5 V_{DIM} . V_{FB} is almost constant with same T_{OFF} in the CCM region.
- <u>B:</u> V_{CS.LIM} is clamped to 80 mV and doesn't changed by V_{DIM}. T_{OFF} is lengthened for dimming as V_{FB} is decreased. Operating mode is transitioned from CCM to DCM at the boundary of A and B region.
- <u>C</u>: When V_{DIM} is lower than 0.25 V, V_{REF} is set to 0 V and V_{FB} is pulled down to 0.5 V clamping voltage with min. LED current under open loop control. When V_{DIM} is further lower than 0.1/0.075 V, standby is triggered with LED turn-off.

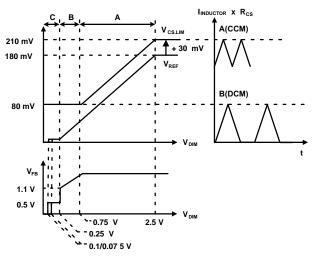


Figure 11. Operation Mode vs. V_{DIM}

Precise CC Regulation

The output of the precise LED current calculator, V_{LED} , is generated by analog sensing amplifiers and V_{LED} is compared with V_{REF} by OTA to generate V_{FB} . Those sensing amplifiers and OTA have zero input offset compensation technique which performs the excellent CC regulation.

Table 1 shows CC tolerance measured by changing inductor ($\pm 20\%$), temperature (-10, 25, 90 °C), output voltage (10, 30, 50 V) and controller 150 pcs(3 lot variation) in 60 V input 75 W driver. As a result, CC tolerance with system variables at 1% deep dimming condition is less than $\pm 20\%$ and less than $\pm 2.0\%$ at full load condition.

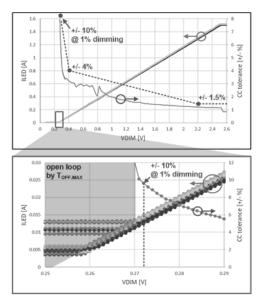


Figure 12. NCL35076 Dimming Curve and CC Tolerance

| Inductor : ± 20% | | | | | | |
|-----------------------------------|-----------|----------|----------|---------|---------|---------|
| Temp. : –10 / 25 / 90 °C | 100% Load | 50% Load | 10% Load | 5% Load | 2% Load | 1% Load |
| V _{OUT} : 10 V | 1.52 | 1.92 | 3.24 | 4.18 | 7.43 | 13.06 |
| V _{OUT} : 30 V | 1.40 | 1.54 | 3.26 | 4.38 | 7.99 | 13.87 |
| V _{OUT} : 50 V | 1.25 | 1.37 | 2.68 | 3.57 | 7.17 | 14.06 |
| V _{OUT} : 10 / 30 / 50 V | 1.62 | 1.98 | 4.10 | 4.94 | 8.24 | 15.40 |

Table 1. CC TOLERANCE (150 pcs)

Standby Mode

Standby mode is triggered by V_{DIM} as shown in Figure 13.

- <u>A:</u> When V_{DIM} is lower than V_{DIM(SB-ENA)}, DRV block is shut down. So, LED lamps turn off.
- <u>B:</u> After t_{SB(DELAY)} (10 ms), standby mode is entered and NCL35076 current consumption drops to I_{DD(SB)}.
- <u>C:</u> When V_{DIM} is higher than V_{DIM(SB-DIS)}, standby mode is immediately terminated and IC starts up.

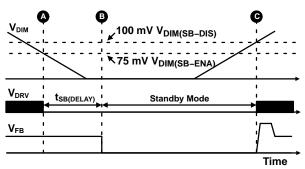
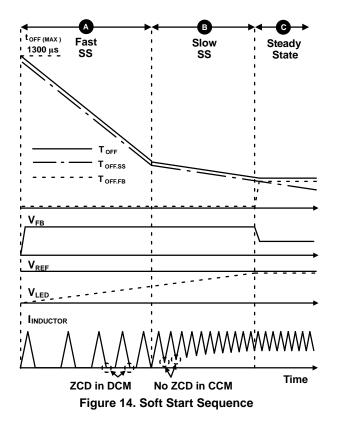


Figure 13. NCL35076 Standby Mode

Soft Start

During soft start operation, T_{OFF} is decided by either T_{OFF_SS} or T_{OFF_FB} . T_{OFF} is governed by T_{OFF_SS} in early start up because T_{OFF_SS} decreases from $t_{OFF(MAX)}$. When T_{OFF_SS} reaches to the steady state level, V_{FB} is settled to the regulation level and T_{OFF} is finally decided by T_{OFF_FB} . In the end of the soft start time, T_{OFF_SS} reaches to 0 and doesn't affect T_{OFF} control anymore. Figure 14 shows how the soft start operates.



- <u>A:</u> V_{FB} is pulled up as V_{LED} is far below V_{REF} . T_{OFF_SS} is reduced quickly from $t_{OFF(MAX)}$ in Fast SS. Fast SS ends when inductor current zero cross (ZCD) is not detected.
- <u>B:</u> Slow SS starts when there is no ZCD in CCM.
- <u>C:</u> V_{LED} is closer to V_{REF}, and V_{FB} starts falling. Then, T_{OFF} is determined by T_{OFF} _{FB} and the steady state starts.

Protections

When protection is triggered, all functional blocks stop operating and begin to start up after 1 second AR time.

- VDD Over Voltage Protection (OVP) When VDD is higher than V_{DD(OVP)} (23 V), VDD OVP is triggered. Open LED protection can be implemented by VDD OVP when VDD is supplied by auxiliary winding in the buck inductor.
- Over Current Protection (OCP)

When CSZCD voltage is higher than $V_{CS(OCP)}$ (0.5 V) after leading edge blanking time, $t_{LEB(TON)}$ (400 ns), IC immediately shuts down.

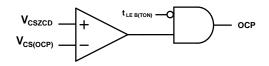


Figure 15. OCP Block

• Short LED Protection (SLP)

When LED load is short–circuited, demagnetizing time of the inductor is very long due to zero output voltage so that T_{OFF} is lengthened and T_{ON} is very short. If CCM and $t_{OFF(MAX)}$ are detected for SLP monitoring time, $t_{SLP(MON-DEL)}$ (20 ms), SLP is triggered. In order to prevent abnormal SLP triggering at startup, SLP monitoring is disabled for $t_{SLP(MON-DIS)}$ (12 ms) after 1st switching begins.

• Thermal Shut Down (TSD)

When the junction temperature is higher than T_{SD} , the system shuts down and the junction temperature is monitored at every 1 second delay time (AR time). When the temperature is lower than $T_{SD} - T_{SD(HYS)}$, the system restarts.

APPENDIX: DIMMING CURVE AND CC TOLERANCE WITH SYSTEM VARIABLES

- System: NCL35076 75 W (V_{IN}: 60V / V_{OUT}: 10 ~ 50V / I_{OUT(MAX)}: 1.5 A)
- Temperature variation: –10 / 25 / 90 $^{\circ}\mathrm{C}$
- Inductance variation: $\pm 20\%$ (120 uH ~ 180 uH)
- Output Voltage: $10\slasharpi$ / $30\slasharpi$ / $50\slasharpi$ V
- NCL35076 Controller: 150 pcs (3 lot variation)

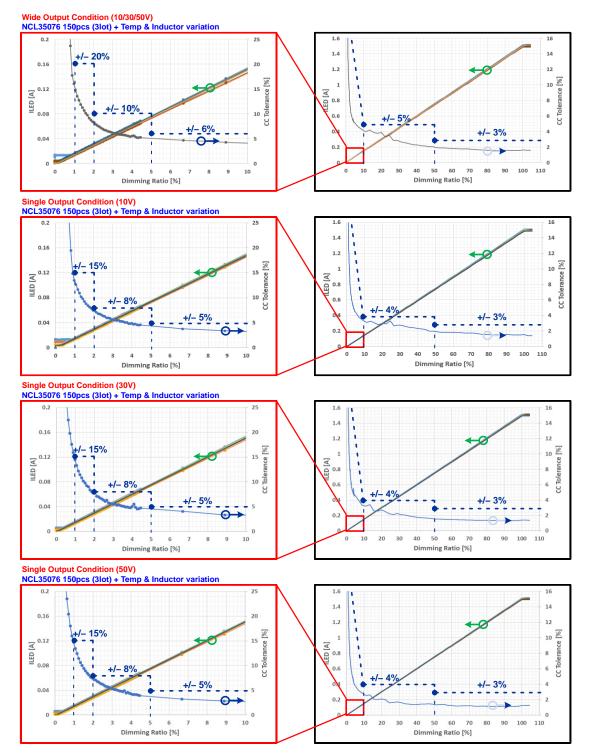
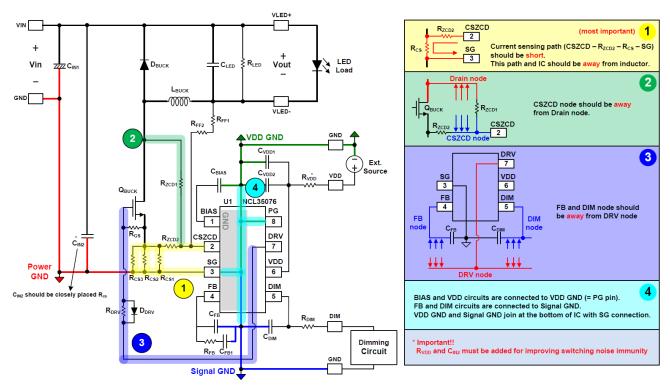


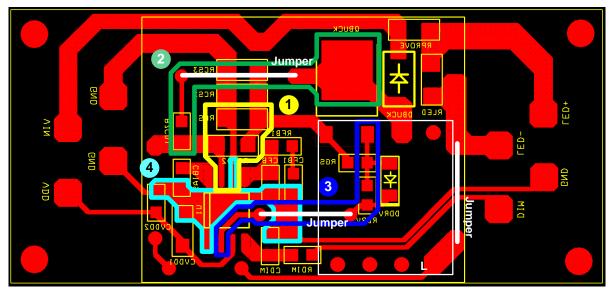
Figure 16. CC Tolerance (150 pcs)

PCB LAYOUT GUIDANCE



(75–W Demo Board Schematic)

(PCB Layout Guidance)



(75-W Demo Board PCB Layout - Bottom)

Figure 17. Layout Guidance

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*For additional information on our Pb-Free strategy and soldering details, please download the **onsemi** Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

STYLES ON PAGE 2

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SOIC-8 NB CASE 751-07 ISSUE AK

STYLE 1: PIN 1. EMITTER COLLECTOR 2. COLLECTOR 3. 4. EMITTER 5. EMITTER BASE 6. 7 BASE EMITTER 8. STYLE 5: PIN 1. DRAIN 2. DRAIN З. DRAIN DRAIN 4. GATE 5. 6. GATE SOURCE 7. 8. SOURCE STYLE 9: PIN 1. EMITTER, COMMON COLLECTOR, DIE #1 COLLECTOR, DIE #2 2. З. EMITTER, COMMON 4. 5. EMITTER, COMMON 6 BASE. DIE #2 BASE, DIE #1 7. 8. EMITTER, COMMON STYLE 13: PIN 1. N.C. 2. SOURCE 3 GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 17: PIN 1. VCC 2. V2OUT V10UT З. TXE 4. 5. RXE 6. VFF 7. GND 8. ACC STYLE 21: PIN 1. CATHODE 1 2. CATHODE 2 3 CATHODE 3 CATHODE 4 4. 5. CATHODE 5 6. COMMON ANODE COMMON ANODE 7. 8. CATHODE 6 STYLE 25: PIN 1. VIN 2 N/C REXT З. 4. GND 5. IOUT 6. IOUT IOUT 7. 8. IOUT STYLE 29: BASE, DIE #1 PIN 1. 2 EMITTER, #1 BASE, #2 З. EMITTER, #2 4. 5 COLLECTOR, #2 COLLECTOR, #2 6.

STYLE 2: PIN 1. COLLECTOR, DIE, #1 2. COLLECTOR, #1 COLLECTOR, #2 3. 4 COLLECTOR, #2 BASE, #2 5. EMITTER, #2 6. 7 BASE #1 EMITTER, #1 8. STYLE 6: PIN 1. SOURCE 2. DRAIN 3. DRAIN SOURCE 4. SOURCE 5. 6. GATE GATE 7. 8. SOURCE STYLE 10: GROUND PIN 1. BIAS 1 OUTPUT 2. З. GROUND 4. 5. GROUND 6 BIAS 2 INPUT 7. 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE 3 P-SOURCE P-GATE 4. P-DRAIN 5 6. P-DRAIN N-DRAIN 7. N-DRAIN 8. STYLE 18: PIN 1. ANODE ANODE 2. SOURCE 3. GATE 4. 5. DRAIN 6 DRAIN CATHODE 7. 8. CATHODE STYLE 22: PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC 3 COMMON CATHODE/VCC 4. I/O LINE 3 COMMON ANODE/GND 5. 6. I/O LINE 4 7. I/O LINE 5 8. COMMON ANODE/GND STYLE 26: PIN 1. GND 2 dv/dt З. ENABLE 4. ILIMIT 5. SOURCE SOURCE 6. SOURCE 7. 8. VCC STYLE 30: DRAIN 1 PIN 1. DRAIN 1 2 GATE 2 З. SOURCE 2 4. SOURCE 1/DRAIN 2 SOURCE 1/DRAIN 2 5. 6.

STYLE 3: PIN 1. DRAIN, DIE #1 DRAIN, #1 2. DRAIN, #2 З. DRAIN, #2 4. GATE, #2 5. SOURCE, #2 6. 7 GATE #1 8. SOURCE, #1 STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS THIRD STAGE SOURCE GROUND З. 4. 5. DRAIN 6. GATE 3 SECOND STAGE Vd 7. FIRST STAGE Vd 8. STYLE 11: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. DRAIN 2 DRAIN 1 7. 8. DRAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 ANODE 1 3 ANODE 1 4. 5. CATHODE, COMMON CATHODE, COMMON CATHODE, COMMON 6. 7. CATHODE, COMMON 8. STYLE 19: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 8. **MIRROR 1** STYLE 23: PIN 1. LINE 1 IN COMMON ANODE/GND COMMON ANODE/GND 2. 3 LINE 2 IN 4. LINE 2 OUT 5. COMMON ANODE/GND COMMON ANODE/GND 6. 7. 8. LINE 1 OUT STYLE 27: PIN 1. ILIMIT 2 OVI 0 UVLO З. 4. INPUT+ 5. 6. SOURCE SOURCE SOURCE 7. 8 DRAIN

STYLE 4: PIN 1. 2. ANODE ANODE ANODE З. 4. ANODE ANODE 5. 6. ANODE 7 ANODE COMMON CATHODE 8. STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 BASE #2 З. COLLECTOR, #2 4. COLLECTOR, #2 5. 6. EMITTER, #2 EMITTER, #1 7. 8. COLLECTOR, #1 STYLE 12: PIN 1. SOURCE SOURCE 2. 3. GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 16 EMITTER, DIE #1 PIN 1. 2. BASE, DIE #1 EMITTER, DIE #2 3 BASE, DIE #2 4. 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 COLLECTOR, DIE #1 7. COLLECTOR, DIE #1 8. STYLE 20: PIN 1. SOURCE (N) GATE (N) SOURCE (P) 2. 3. 4. GATE (P) 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 24: PIN 1. BASE EMITTER 2. 3 COLLECTOR/ANODE COLLECTOR/ANODE 4. 5. CATHODE

6. CATHODE COLLECTOR/ANODE 7. 8. COLLECTOR/ANODE STYLE 28: PIN 1. SW_TO_GND 2. DASIC OFF DASIC_SW_DET З. 4. GND 5. 6. V MON VBULK 7. VBULK

7. VOULK 8. VIN

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SOURCE 1/DRAIN 2

7.

8. GATE 1

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7.

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COLLECTOR, #1

COLLECTOR, #1

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