onsemi

UF4C120053B7S

Description

The UF4C120053B7S is a 1200 V, 53 m Ω G4 SiC FET. It is based on a unique "cascode" circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving TO-263-7 package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

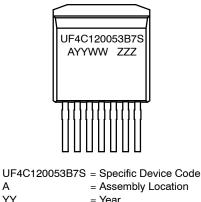
- On-resistance $R_{DS(on)}$: 53 m Ω (Typ)
- Operating Temperature: 175 °C (Max)
- Excellent Reverse Recovery: $Q_{rr} = 98 \text{ nC}$
- Low Body Diode V_{FSD}: 1.28 V
- Low Gate Charge: $Q_G = 37.8 \text{ nC}$
- Threshold Voltage V_{G(th)}: 4.8 V (Typ) Allowing 0 to 15 V Drive
- Low Intrinsic Capacitance
- ESD Protected: HBM Class 2 and CDM Class C3
- TO-263-7 Package for Faster Switching, Clean Gate Waveforms
- This Device is Pb-Free, Halogen Free and is RoHS Compliant

Typical Applications

- EV Charging
- PV Inverters
- Switch Mode Power Supplies
- Power Factor Correction Modules
- Induction Heating

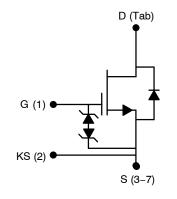
TO-263-7 10.18x9.08x4.43, 1.27P CASE 418BA

MARKING DIAGRAM



A	= Assembly Location
YY	= Year
WW	= Work Week
ZZZ	= Lot ID

PIN CONNECTIONS



ORDERING INFORMATION

See detailed ordering and shipping information on page 8 of this data sheet.

MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Value	Unit
Drain-source Voltage	V _{DS}		1200	V
Gate-source Voltage	V _{GS}	DC	-20 to +20	V
		AC (f > 1 Hz)	–25 to +25	V
Continuous Drain Current (Note 1)	I _D	T _C = 25 °C	34	А
		T _C = 100 °C	24.6	А
Pulsed Drain Current (Note 2)	I _{DM}	T _C = 25 °C	100	А
Single Pulsed Avalanche Energy (Note 3)	E _{AS}	L = 15 mH, I _{AS} = 2.7 A	54.6	mJ
SiC FET dv/dt Ruggedness	dv/dt	$V_{DS} \le 800 \text{ V}$	150	V/ns
Power Dissipation	P _{tot}	T _C = 25 °C	250	W
Maximum Junction Temperature	T _{J, max}		175	°C
Operating and Storage Temperature	T _J , T _{STG}		–55 to 175	°C
Reflow Soldering Temperature	T _{solder}	Reflow MSL 1	245	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Limited by $T_{J, max}$. 2. Pulse width t_p limited by $T_{J, max}$. 3. Starting $T_J = 25$ °C.

THERMAL CHARACTERISTICS

Parameter	Symbol	Test Conditions	Min	Тур	Мах	Unit
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$		-	0.46	0.60	°C/W

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE – STATIC				-		-
Drain-source Breakdown Voltage	BV _{DS}	$V_{GS} = 0 V$, $I_D = 1 mA$	1200	-	-	V
Total Drain Leakage Current	I _{DSS}	V_{DS} = 1200 V, V_{GS} = 0 V, T _J = 25 °C	-	0.2	50	μΑ
		V _{DS} = 1200 V, V _{GS} = 0 V, T _J = 175°C	-	15	-	
Total Gate Leakage Current	I _{GSS}		-	6	20	μΑ
Drain-source On-resistance	R _{DS(on)}	V_{GS} = 12 V, I_D = 20 A, T_J = 25°C	-	53	67	mΩ
		V_{GS} = 12 V, I _D = 20 A, T _J = 125°C	-	112	-	
		V_{GS} = 12 V, I _D = 20 A, T _J = 175°C	-	159	-	
Gate Threshold Voltage	V _{G(th)}	V _{DS} = 5 V, I _D = 10 mA	4	4.8	6	V
Gate Resistance	R _G	f = 1 MHz, open drain	-	4.5	-	Ω

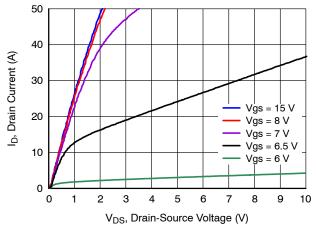
Diode Continuous Forward Current (Note 1) T_C = 25 °C 34 I_S А _ _ T_C = 25 °C 100 А Diode Pulse Current (Note 2) _ IS, pulse _ Forward Voltage $V_{GS} = 0 \text{ V}, \text{ I}_{S} = 10 \text{ A}, \text{ T}_{J} = 25 \ ^{\circ}\text{C}$ v V_{FSD} _ 1.28 1.65 $V_{GS} = 0 \ V, \ I_{S} = 10 \ A, \ T_{J} = 175 \ ^{\circ}C$ 1.96 _ _ Reverse Recovery Charge Q_{rr} 98 nC _ _ 15.2 **Reverse Recovery Time** t_{rr} _ _ ns

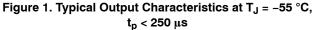
ELECTRICAL CHARACTERISTICS (T_J = +25 $^{\circ}$ C unless otherwise specified) (continued)

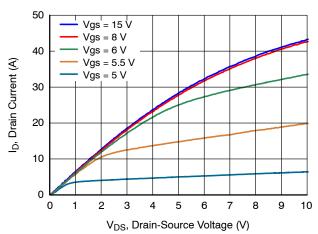
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE - REVERSE DIOD	E	•				
Reverse Recovery Charge	Q _{rr}	V_{DS} = 800 V, I _S = 25 A, V _{GS} = -5 V,	-	105	-	nC
Reverse Recovery Time	t _{rr}	R _G = 20 Ω, di/dt = 1600 A/μs, T _J = 150 °C	-	19.6	-	ns
TYPICAL PERFORMANCE – DYNAMIC		•				
Input Capacitance	C _{iss}	$V_{DS} = 800 \text{ V}, V_{GS} = 0 \text{ V},$	-	1370	-	pF
Output Capacitance	C _{oss}	f = 100 kHz	_	43.5	-	
Reverse Transfer Capacitance	C _{rss}		-	2.2	-	
Effective Output Capacitance, Energy Related	C _{oss(er)}	V_{DS} = 0 V to 800 V, V_{GS} = 0 V	-	54	-	pF
Effective Output Capacitance, Time Related	C _{oss(tr)}		-	100	-	
C _{OSS} Stored Energy	E _{oss}	V_{DS} = 800 V, V_{GS} = 0 V	-	17.3	_	μJ
Total Gate Charge	Q _G	V _{DS} = 800 V, I _D = 25 A,	-	37.8	-	nC
Gate-drain Charge	Q _{GD}	V _{GS} = 0 V to 15 V	-	9.5	-	1
Gate-source Charge	Q _{GS}		-	10	-	
Turn-on Delay Time	t _{d(on)}	$\begin{array}{l} V_{DS}=800 \text{ V}, \text{ I}_{D}=25 \text{ A},\\ \text{Gate Driver}=-5 \text{ V to }+15 \text{ V},\\ \text{R}_{G} \ _{ON}=1 \ \Omega, \ \text{R}_{G} \ _{OFF}=20 \ \Omega,\\ \text{Inductive Load},\\ \text{FWD: Same Device with}\\ \text{V}_{GS}=-5 \text{ V and } \text{R}_{G}=20 \ \Omega,\\ \text{Snubber: } \text{R}_{S}=20 \ \Omega,\\ \text{C}_{S}=100 \ \text{pF}, \end{array}$	-	20	-	ns
Rise Time	tr		-	32	-	-
Turn-off Delay Time	t _{d(off)}		-	57	-	
Fall Time	t _f		-	12	-	
Turn-on Energy Including R _S Energy	E _{ON}		-	570	-	μJ
Turn-off Energy Including R _S Energy	E _{OFF}	T _J = 25 °C	-	57	-	
Total Switching Energy	E _{TOTAL}	(Note 4), (Note 5)	-	627	-	
Snubber R _S Energy During Turn-on	E _{RS_ON}		-	5	-	
Snubber R _S Energy During Turn-off	E_{RS_OFF}		-	11	-	
Turn-on Delay Time	t _{d(on)}	$V_{DS} = 800 \text{ V}, I_{D} = 25 \text{ A},$	-	24	-	ns
Rise Time	t _r	Gate Driver = -5 V to +15 V, R _{G ON} = 1 Ω , R _{G OFF} = 20 Ω ,	-	33	-	
Turn-off Delay Time	t _{d(off)}	Inductive Load, FWD: Same Device with $V_{GS} = -5 V$ and $R_G = 20 \Omega$, Snubber: $R_S = 20 \Omega$, $C_S = 100 \text{ pF}$,	-	63	-	
Fall Time	t _f		-	13	-	
Turn-on Energy Including R _S Energy	E _{ON}		-	660	-	μJ
Turn-off Energy Including R _S Energy	E _{OFF}	T _J = 150 °C	-	75	-	
Total Switching Energy	E _{TOTAL}	(Note 4), (Note 5)	-	735	-	
Snubber R _S Energy During Turn-on	E _{RS_ON}	1	-	5	-	
Snubber R _S Energy During Turn-off	E _{RS_OFF}	1	-	12	-	

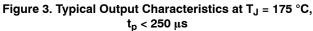
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.
4. Measured with the switching test circuit in Figure 26.
5. In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.

TYPICAL PERFORMANCE DIAGRAMS

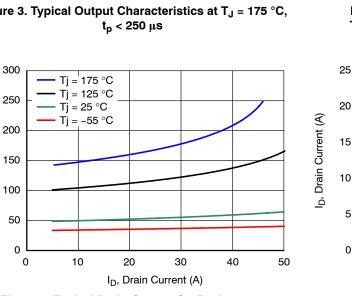


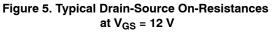


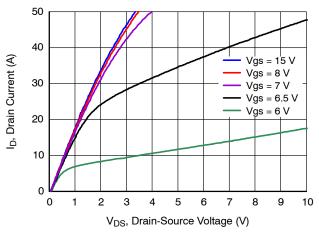


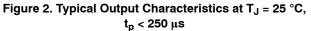


 $R_{DS(on)}$, On-Resistance (m Ω)









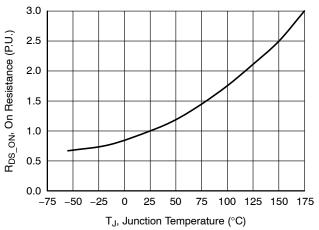
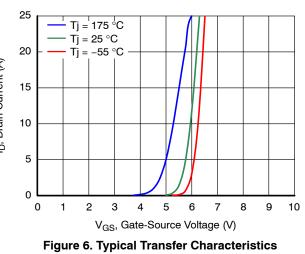


Figure 4. Normalized On-Resistance vs. Temperature at V_{GS} = 12 V and I_D = 25 A



at V_{DS} = 5 V

TYPICAL PERFORMANCE DIAGRAMS (continued)

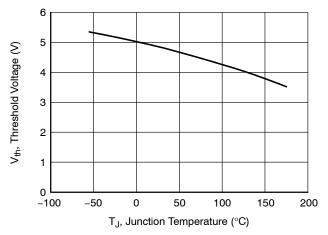


Figure 7. Threshold Voltage vs. Junction Temperature at V_{DS} = 5 V and I_D = 10 mA

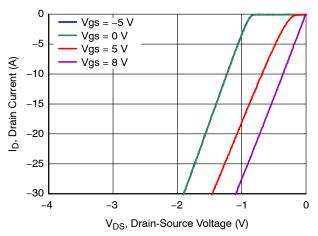


Figure 9. 3rd Quadrant Characteristics at T_J = -55 °C

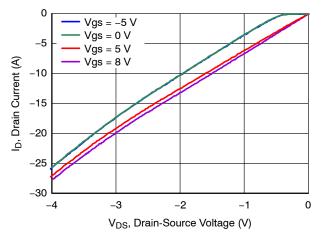
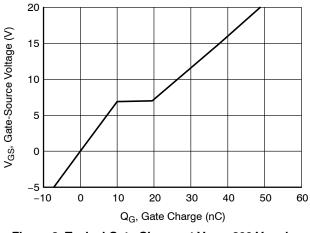
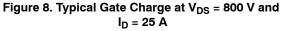


Figure 11. 3rd Quadrant Characteristics at T_J = 175 °C





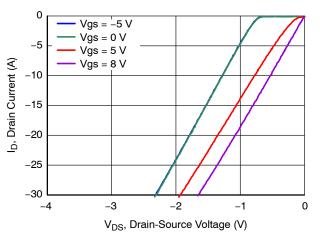


Figure 10. 3rd Quadrant Characteristics at T_J = 25 °C

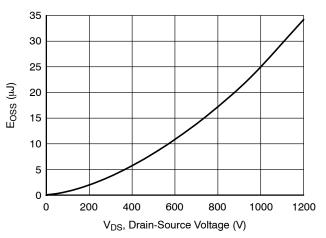
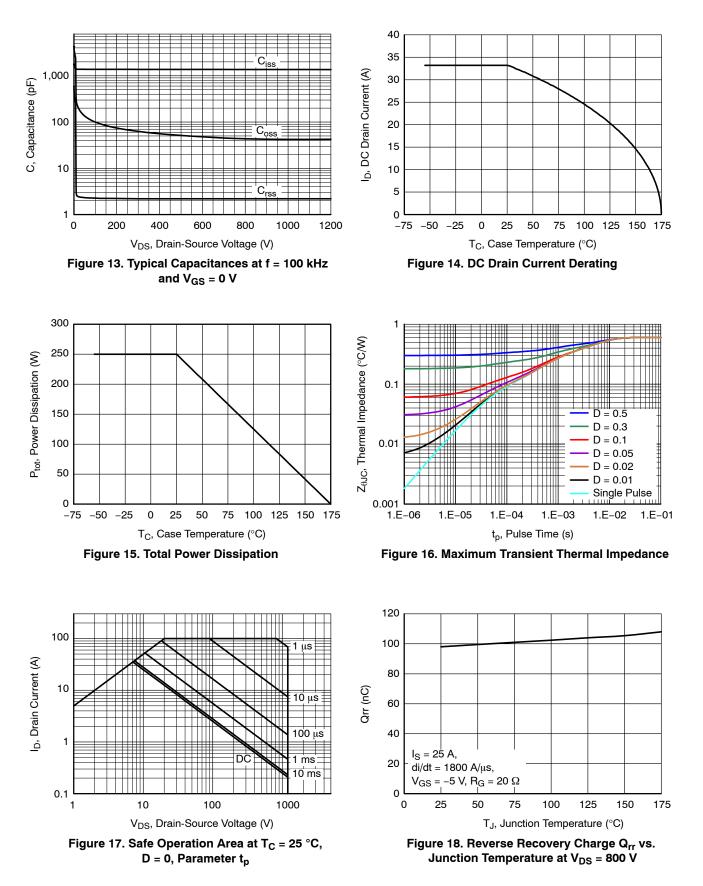
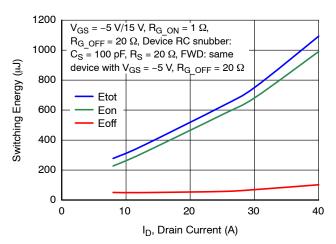


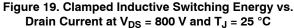
Figure 12. Typical Stored Energy in C_{OSS} at V_{GS} = 0 V

TYPICAL PERFORMANCE DIAGRAMS (continued)



TYPICAL PERFORMANCE DIAGRAMS (continued)





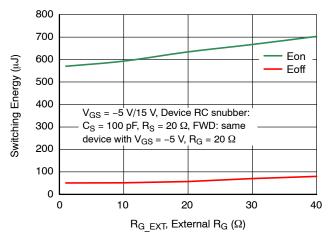
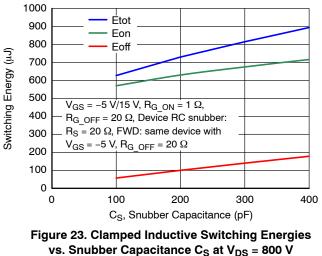


Figure 21. Clamped Inductive Switching Energies vs. $R_{G,EXT}$ at V_{DS} = 800 V, I_D = 25 A, and T_J = 25 °C



 $I_D = 25 \text{ A}, \text{ and } T_J = 25 ^{\circ}\text{C}$

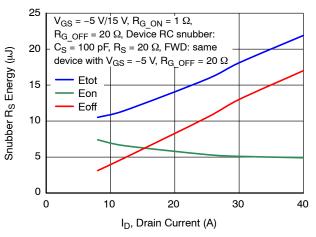


Figure 20. RC Snubber Energy Loss vs. $R_{G,EXT}$ at V_{DS} = 800 V, I_D = 25 A, and T_J = 25 °C

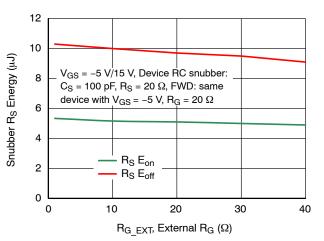


Figure 22. RC Snubber Energy Loss vs. $R_{G,EXT}$ at V_{DS} = 800 V, I_D = 25 A, and T_J = 25 °C

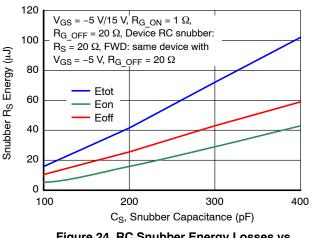


Figure 24. RC Snubber Energy Losses vs. Snubber Capacitance C_S at V_{DS} = 800 V, I_D = 25 A, and T_J = 25 °C

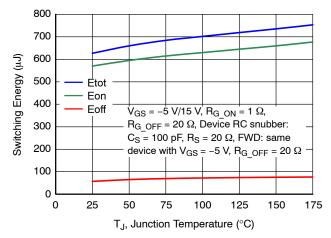


Figure 25. Clamped Inductive Switching Energy vs. Junction Temperature at V_{DS} = 800 V and I_D = 25 A

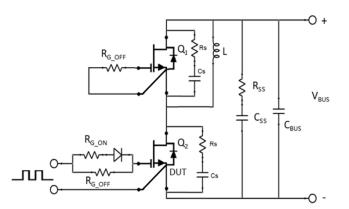


Figure 26. Schematic of the Half-bridge Mode Switching Test Circuit with Device R_C Snubbers (R_S =20 Ω , C_S = 100 pF) and a Bus RC Snubber (R_{SS} = 2.5 Ω , C_{SS} =100 nF)

APPLICATIONS INFORMATION

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

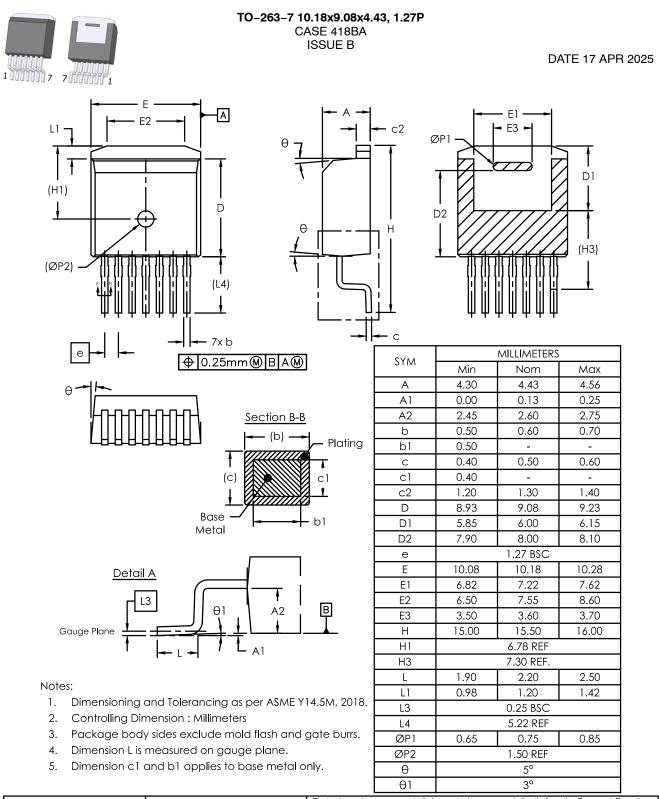
Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see <u>www.onsemi.com</u>. A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the **onsemi** website at www.onsemi.com

ORDERING INFORMATION

Part Number	Marking	Package	Shipping [†]
UF4C120053B7S	UF4C120053B7S	TO-263-7	800 / Tape & Reel

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, <u>BRD8011/D</u>.

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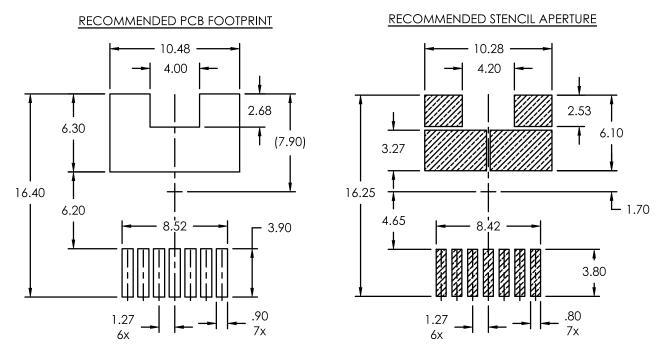


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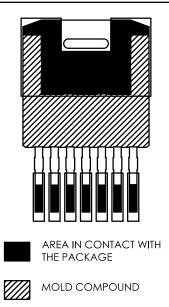
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NOTE: LAND PATTERN AND STENCIL APERTURE DIMENSIONS SERVE ONLY AS AN INITIAL GUIDE. END-USER PCB DESIGN RULES AND TOLERANCES SHOULD ALWAYS PREVAIL.



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