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## AN-8023

# 使用 FAN8303 降壓穩壓器實現負電壓管理

### 摘要

FAN8303 是具有內建功率電晶體(MOSFET)的 2A、370kHz 單晶片型整合降壓穩壓器。它使用簡單,並且只需要最少的外部元件。此技術應用文件將說明如何使用 FAN8303 來產生負電壓。文中將介紹應用範例,並討論升降壓電路的最佳設計。

### 簡介

降壓穩壓器廣泛用於高電壓至低電壓的直流(DC)轉換。同樣地,FAN8303原本的設計是爲了需要調整DC電壓準位的設備上,例如應用在電腦螢幕和電視的機上

盒的微控制器,以及高效能前級線性穩壓器。在某些情況下,非同步降壓穩壓器也可以用於升降壓電路,藉以產生相對於接地電位的負電壓。這些應用包括音頻擴大器和 LCD 面板的定時控制電路等等。

圖 1所示為 LCD 面板的實際應用;LCD 面板需要負電壓進行對比控制。在這個電路方塊中,通常會使用充電泵浦,因為其設計簡單且成本低廉。但相對於輸入電壓的變化,它會有一定數量的功率耗損及較差的輸出電壓調整。具有負輸出的 FAN8303 將會是克服這些問題的解決方案。

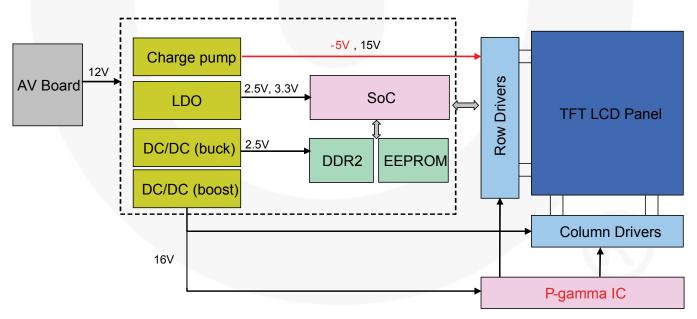


圖 1. 定時控制流程範例

#### 工作原理

爲瞭解升降壓拓樸結構,以下將簡單比較降壓拓樸結構。當 MOSFET 開關(圖 3 中的 Q1)開啟時,電感器( $V_L$ )兩端的電壓差爲  $V_{IN}$ - $V_{OUT}$ 。當 Q1 關閉期間,降壓拓樸結構中的  $V_L$  等於 - $V_{OUT}$ 。所以電感器電流( $I_L$ )會依 ( $V_{IN}$ - $V_{OUT}$ )/L 斜率上升,並依  $V_{OUT}$ /L 斜率下降。因此,電感器所儲存的能量會被轉換成具有正輸出的電壓。此時,若把電感器和飛輪二極體位置交換,則此降壓拓樸結構會變成升降壓拓樸結構。當 MOSFET 開關

Q1(圖 2)開啓時, $V_L$ 與  $V_{IN}$ 相同,因此  $I_L$ 會依  $V_{IN}$ /L 斜率上升。在 Q1 關閉期間, $V_L$  具相反極性,以維持連續電感器電流。因此,它可以產生負輸出電壓- $V_{OUT}$ 。

在設計具降壓穩壓器的升降壓電路時,有幾點需要考量。表1 概略比較降壓與升降壓電路之間的設計參數。

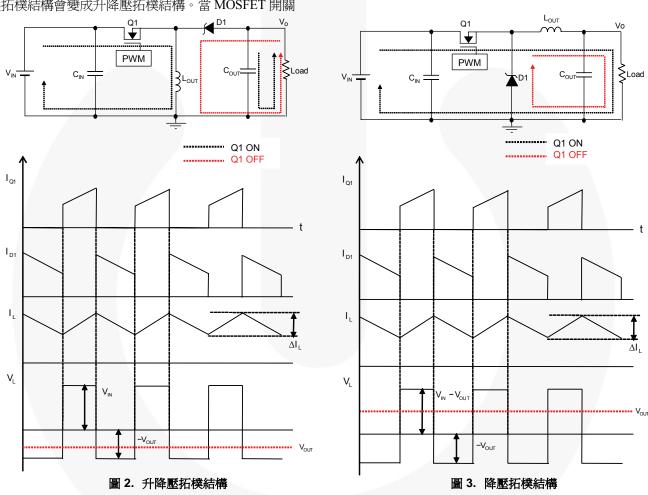


表 1. 降壓與升降壓設計參數

| 拓樸結構 | I <sub>L</sub> (平均值)  | 最大 V <sub>sw</sub>                                | 工作週期  |
|------|-----------------------|---|---|
| 升降壓  | $\frac{I_{OUT}}{1-D}$ | $V_{\mathit{IN}} + \left V_{\mathit{OUT}}\right $ | $\frac{\left V_{\scriptscriptstyle OUT}\right }{V_{\scriptscriptstyle IN} + \left V_{\scriptscriptstyle OUT}\right }$ |
| 降壓   | $I_{OUT}$             | $V_{{\scriptscriptstyle IN}}$                     | $rac{V_{OUT}}{V_{IN}}$   |

首先,電感器電流受限於(1-D);所以請注意,降壓穩壓器的最大輸出電流,一定低於升降壓電路的最大電流。其次,開關節點Vsw為升降壓電路中,輸入電壓與輸出電壓的總和。它亦必須受到降壓穩壓器的最大開關

節點電壓的限制。由於與降壓電路比較起來,升降壓的輸入和輸出有許多雜訊,因此它需要高品質的 MLCC 作 爲輸入和輸出濾波器。

### 設計考量

### 選擇電感器

在選擇電感器時,主要需考量到電感值、RMS 額定電流和 DCR。所採用的電感值通常高於最小電感,以便在「連續電流模式」Continuous Current Mode (CCM)中作業。RMS 電流應高於電感器電流,以防止沒有磁損的電感器飽和。當功率系統需要高效能時,通常會採用低 DCR 的電感器。

若要在連續電流模式中作業,請依下列方程式算出臨界最小電感量:

$$L = \frac{V_{IN} \times D}{f_{SW} \times \Delta I_L} \tag{1}$$

其中:

$$D = \frac{|V_{OUT}|}{|V_{OUT}| + V_{IN}} = 工作週期;$$

f<sub>sw</sub> = 切換頻率;而

### 輸出電容器

為滿足輸出電壓的漣波需求,並在動態負載的情況下維持恆定的輸出電壓,需要使用輸出電容器。漣波電壓會隨 ESR、輸出電容量和 ESL 改變。若要獲得所需的輸出 漣波,下列方程式非常有用,可用來計算所需的最小電容量:

$$C_{MIN} = \frac{I_{OUTMAX} \times D_{MAX}}{f_{SW} \times \Delta V_{OUT}} \tag{2}$$

其中:

 $D_{MAX}$  = 最大工作週期;

I<sub>OUTMAX</sub> = 最大輸出電流;而

 $\Delta V_{OUT}$  = 所需的輸出漣波電壓。

計算所需 ESR 的方程式為:

$$ESR = \frac{\Delta V_{OUT}}{I_{LMAX}}$$
 (3)

### 輸入電容器

輸入電容器應該要能處理最大輸入 RMS 電流,因此請使用下列方程式進行計算。以基層陶瓷電容(MLCC) 爲例,合理的估計值爲每安培(amp)10μF 或 22μF。

最大 RMS 輸入電流:

$$I_{RMS\_MAX} = I_{OUTMAX} \times \sqrt{(D \times (1-D))}$$
 (4)

所需的最小電容量:

$$C_{MIN} = (I_{RMS} \times D) / (f_{SW} \times \Delta V_{IN})$$
 (5)

其中  $\Delta V_{\rm IN}$  是所需的輸入褲波電壓。

#### 飛輪二極體

飛輪二極體的作用,是在開關關閉時,作爲電感器電流放電的路徑。要求低功率耗損時,需考慮崩潰電壓、較低的順向壓降和最大額定電流。此處最好使用蕭基(Schottky)二極體,因其擁有低順向壓降。

所需的二極體額定電流:

$$> I_{LMAX}$$
 (6)

此處的 I<sub>IMAX</sub> 是最大電感器電流。

所需的崩潰電壓:

$$> V_{IN} + |V_{OUT}|$$
 (7)

### 設計範例

以下所示爲設計範例,其測試條件爲: $V_{IN}$  =12V、 $V_{OUT}$  = -5V、 $I_{OUT}$  = 1A 和  $f_{SW}$  =370 kHz(固定)。第一個步驟是設定臨界設計參數,例如電感器漣波電流( $\Delta I_L$ )和所需的輸出漣波電壓( $\Delta V_{OUT}$ )。第二個步驟是計算工作週期。爲得出正確的值,需將二極體的順向壓降和MOSFET 導通時的壓降納入考慮。Fairchild FAN8303是非同步的降壓穩壓器,它整合了  $0.22\Omega$  的 N 通道

MOSFET,因此壓降約為 0.4V。蕭基(Schottky)二極 體的順向電壓( $40V_{RRM}/2A~I_{OUT}$ )為 0.45V。而關於電 感器的部分,建議您使用高於計算值的電感值,並且最 好使用低 DCR 電感器:

#### 表 2. 設計範例計算

| 工作週期:    | $= ( V_{OUT} +V_F) / (V_{IN}+ V_{OUT} +V_F-V_{Q1})$  | 0.33                                 |
|----------|--|--------------------------------------|
| 電感量:     | $= (V_{\rm IN} \times D) / (f_{\rm SW} \times \Delta I_{\rm L})$                               | 35.6μH (所需的 $\Delta I_L$ = 20%)      |
| 輸出電容量:   | $= (I_{OUT} \times D) / (f_{SW} \times \Delta V_{OUT})$  | 86.8μF(所需的 ΔV <sub>OUT</sub> = 10mV) |
| 輸入電容量:   | $I_{RMS} = I_{OUT} \times \sqrt{D \times (1 - D)}$   | 0,47A                                |
|          | $C_{IN} = I_{RMS} \times D / (\Delta V_{IN} \times f_{SW})$                                    | 4.05μF                               |
| 二極體額定電流: | $I_{\text{DIODE\_MAX}} = I_{\text{AVG}} + \Delta I_{\text{L}}/2$ 此處 $I_{\text{AVG}} =$ 平均電感器電流 | 1.77A                                |

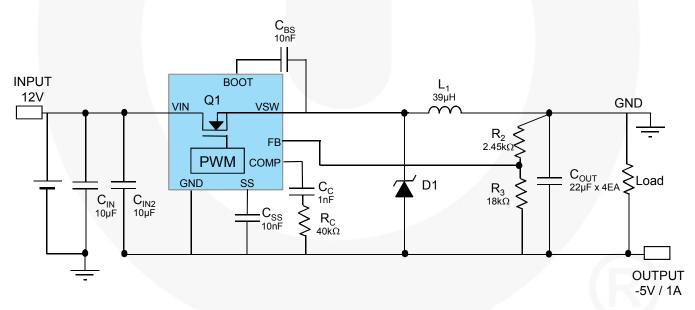
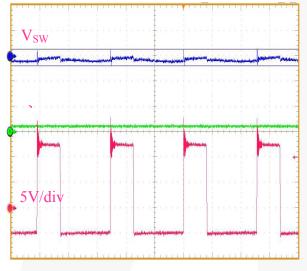


圖 4. 使用 FAN8303 的升降壓電路圖

### 一般波形和圖表

圖 5 和 圖 6 所示為 FAN8303 輸出漣波電壓的一般波 形。為達到低漣波電壓,需使用低於  $10m\Omega$  的 MLCC。

圖 7 所示為 FAN8303 效率與功耗圖。圖中表示,當負載條件為 400mA、功耗為 0.31W 的情況下,最高效率 為 87%。



Vout SomV/div

Iout SomA/div

Vsw 5V/div

圖 5. 33mV 在 100mA 時的 V<sub>OUT</sub> 漣波(1μs/div)

圖 6. 89mV 在 1A 時的 V<sub>OUT</sub> 漣波 (1μs/div)

#### 附註:

1. 測試條件:  $V_{IN}$  =12V 、 $V_{OUT}$  = -5V 、 $f_{SW}$  = 固定 370 kHz 和  $I_{OUT}$  = 0~1A 。



圖 7. 效率與功耗

### 結論

Fairchild 2A 單晶片型,非同步降壓穩壓器 FAN8303 的輸入範圍廣(~23V),並且有絕佳的負載和和線性調整率。儘管是降壓穩壓器拓樸架構,但只要變更被動元

件,便可以在升降壓電路中使用 FAN8303 來產生負輸 出電壓。

### 相關資料表

FAN8303 — 2A 23V 非同步降壓式 DC/DC 穩壓器

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